

# MAXIMUM POWER POINT TRACKING BASED HYBRID WINDSOLAR ENERGY SYSTEM

**Kaladhar Gunnala<sup>1</sup>, D.BALA GANGI REDDY<sup>2</sup>, DR.SUKHDEO SAO<sup>3</sup>**

<sup>1</sup>M-tech Student Scholar, <sup>2</sup>Associate Professor, (Ph.d), <sup>3</sup>Professor, Ph.d Department of Electrical & Electronics Engineering, Bharat Institute of Engineering and technology, Mangalpally; Ranga Reddy (Dt); A.P, India

**Abstract**— Environmentally friendly solutions are becoming more prominent than ever as a result of concern regarding the state of our deteriorating planet. This paper presents a new system configuration of the front-end rectifier stage for a hybrid wind/photovoltaic energy system. This configuration allows the two sources to supply the load separately or simultaneously depending on the availability of the energy sources. The inherent nature of this Cuk-SEPIC fused converter, additional input filters are not necessary to filter out high frequency harmonics. Harmonic content is detrimental for the generator lifespan, heating issues, and efficiency. The fused multi input rectifier stage also allows Maximum Power Point Tracking (MPPT) to be used to extract maximum power from the wind and sun when it is available. An adaptive MPPT algorithm will be used for the wind system and a standard perturb and observe method will be used for the PV system. Operational analysis of the proposed system will be discussed in this paper. Simulation results are given to highlight the merits of the proposed circuit.

**Index Terms**— Maximum Power Point Tracking (MPPT), Cuk-SEPIC fused converter

## I. INTRODUCTION

With increasing concern of global warming and the depletion of fossil fuel reserves, many are looking at sustainable energy solutions to preserve the earth for the future generations. Other than hydro power, wind and photovoltaic energy holds the most potential to meet our energy demands. Alone, wind energy is capable of supplying large amounts of power but its presence is highly unpredictable as it can be here one moment and gone in another. Similarly, solar energy is present throughout the day but the solar irradiation levels vary due to sun intensity and unpredictable shadows cast by clouds, birds, trees, etc. The common inherent drawback of wind and photovoltaic systems are their intermittent natures that make them unreliable. However, by combining these two intermittent sources and by

incorporating maximum power point tracking(MPPT) algorithms, the system's power transfer efficiency and reliability can be improved significantly.

When a source is unavailable or insufficient in meeting the load demands, the other energy source can compensate for the difference. Several hybrid wind/PV power systems with MPPT control have been proposed and discussed in works [1]- [5]. Most of the systems in literature use a separate DC/DC boost converter connected in parallel in the rectifier stage as shown in Figure 1 to perform the MPPT control for each of the renewable energy power sources [1]-[4].

A simpler multi input structure has been suggested by [5] that combine the sources from the DC-end while still achieving MPPT for each renewable source. The structure proposed by [5] is a fusion of the buck and buck-boost converter. The systems in literature require passive input filters to remove the high frequency current harmonics injected into wind turbine generators [6].

The harmonic content in the generator current decreases its lifespan and increases the power loss due to heating [6]. In this paper, an alternative multi-input rectifier structure is proposed for hybrid wind/solar energy systems. The proposed design is a fusion of the Cuk and SEPIC converters.

The features of the proposed topology are: 1) the inherent nature of these two converters eliminates the need for separate input filters for PFC [7]-[8]; 2) it can support step up/down operations for each renewable source (can support wide ranges of PV and wind input); 3) MPPT can be realized for each source; 4) individual and simultaneous operation is supported.

The circuit operating principles will be discussed in this paper. Simulation results are provided to verify with the feasibility of the proposed system

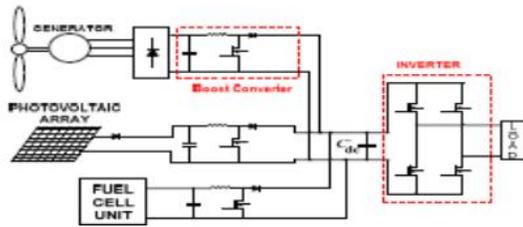


Figure 1: Hybrid system with multi-connected boost converter

## II. PROPOSED MULTI-INPUT RECTIFIER STAGE

A system diagram of the proposed rectifier stage of a hybrid energy system is shown in Figure 2, where one of the inputs is connected to the output of the PV array and the other input connected to the output of a generator. The fusion of the two converters is achieved by reconfiguring the two existing diodes from each converter and the shared utilization of the Cuk output inductor by the SEPIC converter. This configuration allows each converter to operate normally individually in the event that one source is unavailable. Figure 3 illustrates the case when only the wind source is available. In this case, D1 turns off and D2 turns on; the proposed circuit becomes a SEPIC converter and the input to output voltage relationship is given by (1). On the other hand, if only the PV source is available, then D2 turns off and D1 will always be on and the circuit becomes a Cuk converter as shown in Figure 4. The input to output voltage relationship is given by (2). In both cases, both converters have step-up/down capability, which provide more design flexibility in the system if duty ratio control is utilized to perform MPPT control.

$$\frac{V_{dc}}{V_w} = \frac{d_2}{1-d_2} \quad (1)$$

$$\frac{V_{dc}}{V_{PV}} = \frac{d_1}{1-d_1} \quad (2)$$

Figure 5 illustrates the various switching states of the proposed converter. If the turn on duration of M1 is longer than M2, then the switching states will be state I, II, IV. Similarly, the switching states will be state I, III, IV if the switch conduction periods are vice versa. To provide a better explanation, the inductor current waveforms of each switching state are given as follows assuming that  $d_2 > d_1$ ; hence only states I, III, IV are discussed in this example. In the following,  $I_{i,PV}$  is the average input current from the PV source;  $I_{i,W}$  is the RMS input current after the rectifier (wind case); and  $I_{dc}$  is the average system output current. The key waveforms

that illustrate the switching states in this example are shown in Figure 6. The mathematical expression that relates the total output voltage and the two input sources will be illustrated in the next section.

State I ( $M_1$  on,  $M_2$  on):

$$i_{L1} = I_{i,PV} + \frac{V_{PV}}{L_1}t \quad 0 < t < d_1T_s$$

$$i_{L2} = I_{dc} + \left( \frac{v_{c1} + v_{c2}}{L_2} \right)t \quad 0 < t < d_1T_s$$

$$i_{L3} = I_{i,W} + \frac{V_w}{L_3}t \quad 0 < t < d_1T_s$$

State III ( $M_1$  off,  $M_2$  on):

$$i_{L1} = I_{i,PV} + \left( \frac{V_{PV} - v_{c1}}{L_1} \right)t \quad d_1T_s < t < d_2T_s$$

$$i_{L2} = I_{dc} + \frac{v_{c2}}{L_2}t \quad d_1T_s < t < d_2T_s$$

$$i_{L3} = I_{i,W} + \frac{V_w}{L_3}t \quad d_1T_s < t < d_2T_s$$

State IV ( $M_1$  off,  $M_2$  off):

$$i_{L1} = I_{i,PV} + \left( \frac{V_{PV} - v_{c1}}{L_1} \right)t \quad d_2T_s < t < T_s$$

$$i_{L2} = I_{dc} - \frac{V_{dc}}{L_2}t \quad d_2T_s < t < T_s$$

$$i_{L3} = I_{i,W} + \left( \frac{V_w - v_{c2} - V_{dc}}{L_3} \right)t \quad d_2T_s < t < T_s$$

Figure 2: Proposed rectifier stage for a Hybrid wind/PV

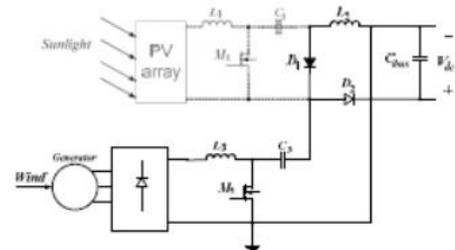


Figure 3: Only wind source is operational (SEPIC)

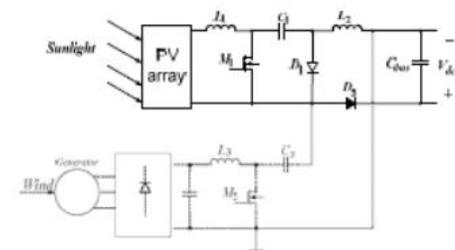


Figure 4: Only PV source is operation (Cuk)

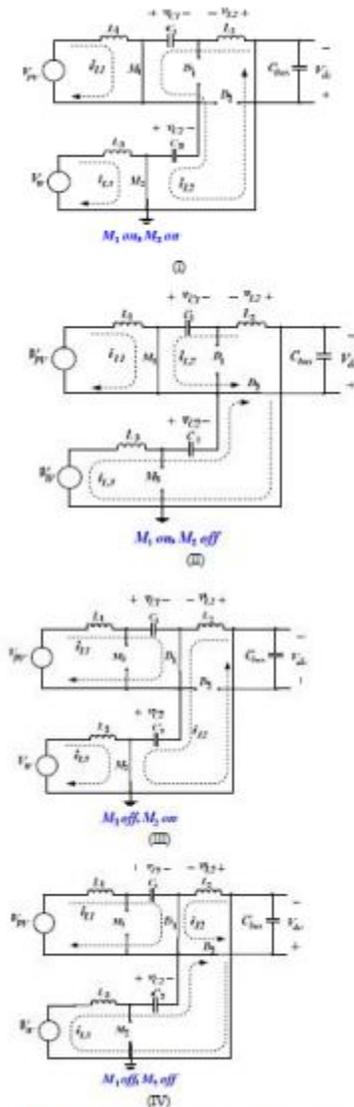


Figure 5 (I-IV): switching states within a switching cycle

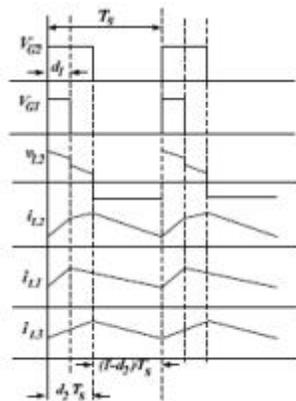


Figure 6: Proposed circuit inductor waveforms

find an expression for the output DC bus voltage,  $V_{dc}$ , the volt-balance of the output inductor,  $L_2$ , is examined according to Figure 6 with  $d_2 > d_1$ . Since the net change in the voltage of  $L_2$  is zero, applying volt-balance to  $L_2$  results in (3). The expression that relates the average output DC voltage ( $V_{dc}$ ) to the capacitor voltages ( $v_{c1}$  and  $v_{c2}$ ) is then obtained as shown in (4), where  $v_{c1}$  and  $v_{c2}$  can then be obtained by applying volt-balance to  $L_1$  and  $L_3$  [9].

The final expression that relates the average output voltage and the two input sources ( $V_W$  and  $V_{PV}$ ) is then given by (5). It is observed that  $V_{dc}$  is simply the sum of the two output voltages of the Cuk and SEPIC converter. This further implies that  $V_{dc}$  can be controlled by  $d_1$  and  $d_2$  individually or simultaneously.

$$(v_{c1} + v_{c2})d_2 I_s + (v_{c2})(d_2 - d_1)I_s + (1 - d_2)(-V_{dc})I_s = 0 \quad (3)$$

$$V_{dc} = \left(\frac{d_1}{1-d_2}\right)v_{c1} + \left(\frac{d_2}{1-d_2}\right)v_{c2} \quad (4)$$

$$V_{dc} = \left(\frac{d}{1-d_1}\right)V_{PV} + \left(\frac{d_2}{1-d_2}\right)V_w \quad (5)$$

The switches voltage and current characteristics are also provided in this section.

The voltage stress is given by (6) and (7) respectively. As for the current stress, it is observed from Figure 6 that the peak current always occurs at the end of the on-time of the MOSFET. Both the Cuk and SEPIC MOSFET current consists of both the input current and the capacitors ( $C_1$  or  $C_2$ ) current.

The peak current stress of  $M_1$  and  $M_2$  are given by (8) and (10) respectively.  $L_{eq1}$  and  $L_{eq2}$ , given by (9) and (11), represent the equivalent inductance of Cuk and SEPIC converter respectively.

The PV output current, which is also equal to the average input current of the Cuk converter is given in (12). It can be observed that the average inductor current is a function of its respective duty cycle ( $d_1$ ).

Therefore by adjusting the respective duty cycles for each energy source, maximum power point tracking can be achieved.

$$v_{a1} = V_w \left( 1 + \frac{d_1}{1-d_1} \right) \quad (6)$$

$$v_{a2} = V_w \left( 1 + \frac{d_2}{1-d_2} \right) \quad (7)$$

$$i_{a1,p1} = I_{1,pr} + I_{a,p1} + \frac{V_w d_1 I_1}{2L_{eq1}} \quad (8)$$

$$L_{eq1} = \frac{L_1 L_2}{L_1 + L_2} \quad (9)$$

$$i_{a1,p2} = I_{1,pr} + I_{a,p2} + \frac{V_w d_2 I_2}{2L_{eq2}} \quad (10)$$

$$L_{eq2} = \frac{L_1 L_2}{L_1 + L_2} \quad (11)$$

$$I_{1,pr} = \frac{P_m d_1}{V_w (1-d_1)} \quad (12)$$

$$\lambda = \frac{R \omega_b}{v_w} \quad (14)$$

Where

R = turbine radius,

$\omega_b$  = angular rotational speed

### III. MPPT CONTROL OF PROPOSED CIRCUIT

A common inherent drawback of wind and PV systems is the intermittent nature of their energy sources. Wind energy is capable of supplying large amounts of power but its presence is highly unpredictable as it can be here one moment and gone in another. Solar energy is present throughout the day, but the solar irradiation levels vary due to sun intensity and unpredictable shadows cast by clouds, birds, trees, etc. These drawbacks tend to make these renewable systems inefficient. However, by incorporating maximum power point tracking (MPPT) algorithms, the systems' power transfer efficiency can be improved significantly. To describe a wind turbine's power characteristic, equation (13) describes the mechanical power that is generated by the wind [6].

$$P_m = 0.5 \rho A C_p (\lambda, \beta) v_w^3 \quad (13)$$

Where

$\rho$  = air density,

A = rotor swept area,

$C_p(\lambda, \beta)$  = power coefficient function

$\lambda$  = tip speed ratio,

$\beta$  = pitch angle,

$v_w$  = wind speed

The power coefficient ( $C_p$ ) is a nonlinear function that represents the efficiency of the wind turbine to convert wind energy into mechanical energy.

It is dependent on two variables, the tip speed ratio (TSR) and the pitch angle. The TSR,  $\lambda$ , refers to a ratio of the turbine angular speed over the wind speed. The mathematical representation of the TSR is given by (14) [10]. The pitch angle,  $\beta$ , refers to the angle in which the turbine blades are aligned with respect to its longitudinal axis.

Figure 7 and 8 are illustrations of a power coefficient curve and power curve for a typical fixed pitch ( $\beta = 0$ ) horizontal axis wind turbine.

It can be seen from figure 7 and 8 that the power curves for each wind speed has a shape similar to that of the power coefficient curve. Because the TSR is a ratio between the turbine rotational speed and the wind speed, it follows that each wind speed would have a different corresponding optimal rotational speed that gives the optimal TSR. For each turbine there is an optimal TSR value that corresponds to a maximum value of the power coefficient ( $C_{p,max}$ ) and therefore the maximum power. Therefore by controlling rotational speed, (by means of adjusting the electrical loading of the turbine generator) maximum power can be obtained for different wind speeds.

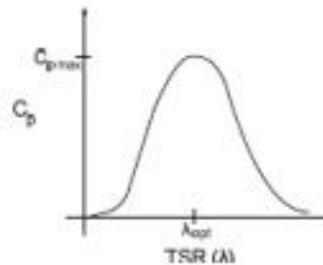


Figure 7: Power Coefficient Curve for a typical wind turbine

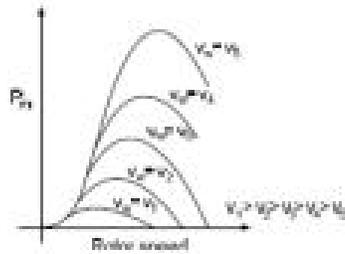


Figure 8: Power Curves for a typical wind turbine

A solar cell is comprised of a P-N junction semiconductor that produces currents via the photovoltaic effect. PV arrays are constructed by placing numerous solar cells connected in series and in parallel [5]. A PV cell is a diode of a large-area forward bias with a photo voltage and the equivalent circuit is shown by Figure 9 [11]. The current-voltage characteristic of a solar cell is derived in [12] and [13] as follows:

$$I = I_{ph} - I_D \quad (15)$$

$$I = I_{ph} - I_0 \left[ \exp\left(\frac{q(V + R_s I)}{A k_B T}\right) - 1 \right] - \frac{V + R_s I}{R_{sh}} \quad (16)$$

Where

- $I_{ph}$  = photocurrent,
- $I_D$  = diode current,
- $I_0$  = saturation current,
- $A$  = ideality factor,
- $q$  = electronic charge  $1.6 \times 10^{-19}$ ,
- $k_B$  = Boltzmann's gas constant  $(1.38 \times 10^{-23})$ ,
- $T$  = cell temperature,
- $R_s$  = series resistance,
- $R_{sh}$  = shunt resistance,
- $I$  = cell current,
- $V$  = cell voltage

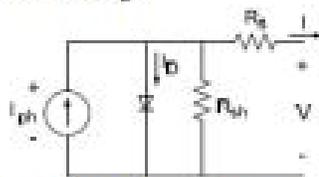


Figure 9: PV cell equivalent circuit

Typically, the shunt resistance ( $R_{sh}$ ) is very large and the series resistance ( $R_s$ ) is very small [5]. Therefore, it is common to neglect these resistances in order to simplify the solar cell model. The resultant ideal voltage-current characteristic of a photovoltaic cell is given by (17) and illustrated by Figure 10. [5]

$$I = I_{ph} - I_0 \left( \exp\left(\frac{qV}{k_B T}\right) - 1 \right) \quad (17)$$

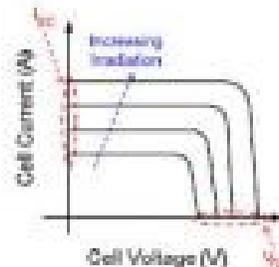


Figure 10: PV cell voltage-current characteristic

The typical output power characteristics of a PV array under various degrees of irradiation is illustrated by Figure 11. It can be observed in Figure 11 that there is a particular optimal voltage for each irradiation level that corresponds to maximum output power. Therefore by adjusting the output current (or voltage) of the PV array, maximum power from the array can be drawn.

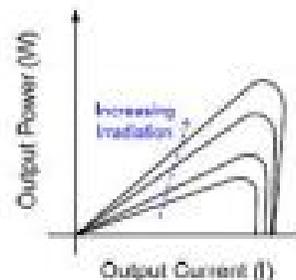


Figure 11: PV cell power characteristics

Due to the similarities of the shape of the wind and PV array power curves, a similar maximum power point tracking scheme known as the hill climb search (HCS) strategy is often applied to these energy sources to extract maximum power. The HCS strategy perturbs the operating point of the system and observes the output.

If the direction of the perturbation (e.g. an increase or decrease in the output voltage of a PV array) results in a positive change in the output power, then the control algorithm will continue in the direction of the previous perturbation. Conversely, if a negative change in the output power is observed, then the control algorithm will reverse the direction of the previous perturbation step. In the case that the change in power is close to zero (within a specified range) then the algorithm will invoke no changes to the system operating point since it corresponds to the

maximum power point (the peak of the power curves). The MPPT scheme employed in this paper is a version of the HCS strategy. Figure 12 is the flow chart that illustrates the implemented MPPT scheme.

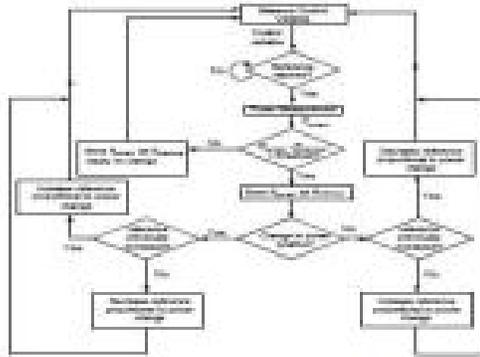
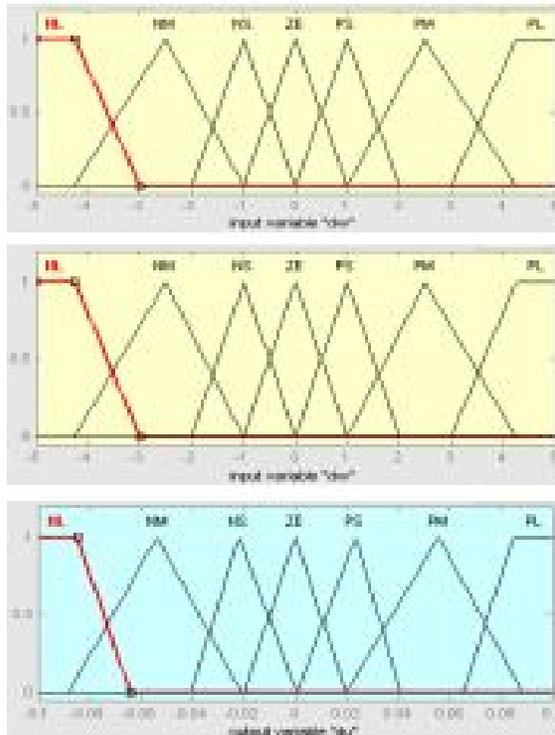


Figure 12: General MPPT Flow Chart for wind and PV

**FUZZY LOGIC**



**RULE table**

(dP/dI)/Iu	NL	NM	NS	ZE	PS	PM	PL
NL	NL	NL	NL	NM	NM	NS	ZE
NM	NL	NL	NM	NM	NS	ZE	PL
NS	NL	NM	NS	NS	ZE	PS	PM
ZE	NL	NM	NS	ZE	PS	PM	PL
PS	NM	NS	ZE	PS	PS	PM	PL
PM	NS	ZE	PS	PM	PM	PL	PL
PL	ZE	PS	PM	PL	PL	PL	PL

In recent years, the number and variety of applications of fuzzy logic have increased significantly. The applications range from consumer products such as cameras, camcorders, washing machines, and microwave ovens to industrial process control, medical instrumentation, decision-support systems, and portfolio selection.

To understand why use of fuzzy logic has grown, you must first understand what is meant by fuzzy logic. Fuzzy logic has two different meanings. In a narrow sense, fuzzy logic is a logical system, which is an extension of multivalve logic. However, in a wider sense fuzzy logic (FL) is almost synonymous with the theory of fuzzy sets, a theory which relates to classes of objects with unsharp boundaries in which membership is a matter of degree. In this perspective, fuzzy logic in its narrow sense is a branch of fl. Even in its more narrow definition, fuzzy logic differs both in concept and substance from traditional multivalve logical systems.

In fuzzy Logic Toolbox software, fuzzy logic should be interpreted as FL, that is, fuzzy logic in its wide sense. The basic ideas underlying FL are explained very clearly and insightfully in Foundations of Fuzzy Logic. What might be added is that the basic concept underlying FL is that of a linguistic variable, that is, a variable whose values are words rather than numbers. In effect, much of FL may be viewed as a methodology for computing with words rather than numbers. Although words are inherently less precise than numbers, their use is closer to human intuition. Furthermore, computing with words exploits the tolerance for imprecision and thereby lowers the cost of solution.

Another basic concept in FL, which plays a central role in most of its applications, is that of a fuzzy if-then rule or, simply, fuzzy rule. Although rule-based systems have a long history of use in Artificial Intelligence (AI), what is missing in such systems is a mechanism for dealing with fuzzy consequents and fuzzy antecedents. In fuzzy logic, this mechanism is provided by the calculus of fuzzy rules. The calculus of fuzzy rules serves as a basis for what might be called the Fuzzy Dependency and Command Language (FDCL).

Although FDCL is not used explicitly in the toolbox, it is effectively one of its principal constituents. In most of the applications of fuzzy logic, a fuzzy logic solution is, in reality, a translation of a human solution into FDCL. A trend that is growing in visibility relates to the use of fuzzy logic in combination with neuron computing and genetic algorithms. More generally, fuzzy logic, neuron computing, and genetic algorithms may be viewed as the principal constituents of what might be called soft computing. Unlike the traditional, hard computing, soft computing accommodates the imprecision of the real world. The guiding principle of soft computing is: Exploit the tolerance for imprecision, uncertainty, and partial truth to achieve tractability, robustness, and low solution cost. In the future, soft computing could play an increasingly important role in the conception and design of systems whose MIQ (Machine IQ) is much higher than that of systems designed by conventional methods.

Among various combinations of methodologies in soft computing, the one that has highest visibility at this juncture is that of fuzzy logic and neuro computing, leading to neuro-fuzzy systems. Within fuzzy logic, such systems play a particularly important role in the induction of rules from observations. An effective method developed by

Dr. Roger Jang for this purpose is called ANFIS (Adaptive Neuro-Fuzzy Inference System). This method is an important component of the toolbox. The fuzzy logic toolbox is highly impressive in all respects. It makes fuzzy logic an effective tool for the conception and design of intelligent systems. The fuzzy logic toolbox is easy to master and convenient to use. And last, but not least important, it provides a reader friendly and up-to-date introduction to methodology of fuzzy logic and its wide ranging applications .

#### IV. SIMULATION RESULTS

In this section, simulation results from PSIM 8.0.7 is given to verify that the proposed multi-input rectifier stage can support individual as well as simultaneous

mode) is supplying power to the load. Figure 14 illustrates the system where only the wind turbine generates power to the load (SEPIC converter mode).

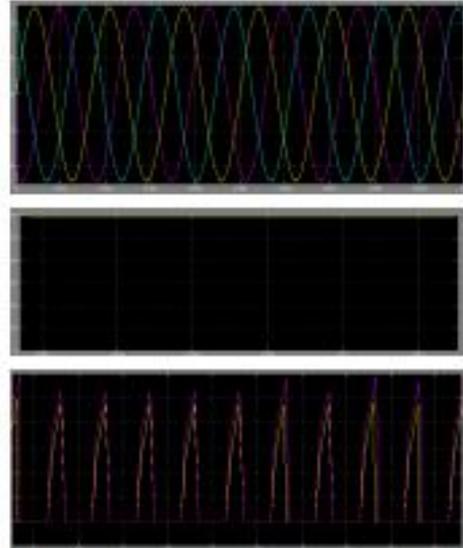


Figure 15 : Simultaneous operation with both wind and PV source (Fusion mode with Cuk and SEPIC) (I) The injected three phase generator current; (II) Top: Output power, Bottom: Switch currents (M1 and M2)

Figure 16 and 17 illustrates the MPPT operation of the PV component of the system (Cuk operation) and the Wind component of the system (SEPIC operation) respectively.



Figure 16 : Solar MPPT – PV output current and reference current signal (Cuk operation)

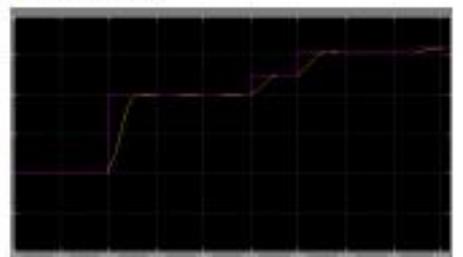


Figure 17 : Wind MPPT – Generator speed and reference speed signal (SEPIC operation)

Finally, Figure 15 illustrates the simultaneous operation (Cuk-SEPIC fusion mode) of the two sources where M2 has a longer conduction cycle (converter states I, IV and III—see Figure 5).

operation. The specifications for the design example are given in TABLE I. Figure 13 illustrates the system under the condition where the wind source has failed and only the PV source (Cuk converter

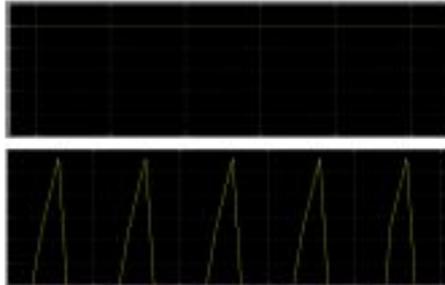


Figure 13 : Individual operation with only PV source (Cuk operation)Top: Output power, Bottom: Switch currents (M1 and M2)

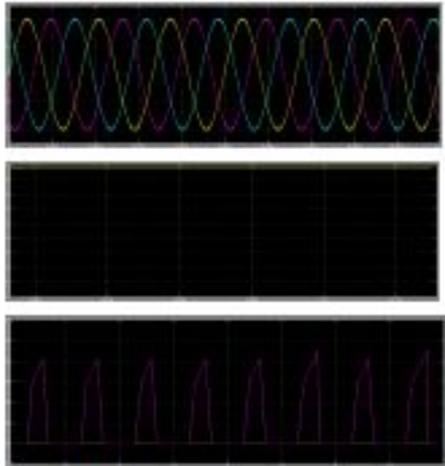


Figure 14 : Individual operation with only wind source (SEPIC operation)(I) The injected three phase generator current; (II) Top: Output power, Bottom: Switch currents (M1 and M2)

Engineering, Hyderabad in 1999 and 2005 respectively. Currently he is an Associate Professor in the Department of Electrical and Electronics Engineering at the Bharat Institute of Engineering and Technology. And Pursuing Ph.D student at Jawaharlal Nehru Technological University, Hyderabad, under the guidance of Dr. M. Suryakalavathi Prof. His research interests include control of electrical drives, Electrical distribution system, power quality and FACTS.



Dr. SUKHDEO SAO presently working as PRINCIPAL in BHARAT Engineering College. He did B.SC ENGG degree in Electrical &

Electronics Engineering from BIT sindri, Dhanbad. And then completed his M.TECH in Electrical & Electronics Engineering as EMED Engineering is specialization at REC WARANGAL. And he done ph.d degree in power systems in BRA Muzaffarpur, Bihar. He has a teaching experience of 30 years.

## V. CONCLUSION

In this paper a new multi-input Cuk-SEPIC rectifier stage for hybrid wind/solar energy systems has been presented. The features of this circuit are: 1) additional input filters are not necessary to filter out high frequency harmonics; 2) both renewable sources can be stepped up/down (supports wide ranges of PV and wind input); 3) MPPT can be realized for each source; 4) individual and simultaneous operation is supported. Simulation results have been presented to verify the features of the proposed topology

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## AUTHORS



Kaladhar.G received B.TECH degree from B.L.E.T college in the year 2011 and currently pursuing M.Tech in Electrical power engineering at BHARAT College Engineering. And his areas of interest are Power Quality control, Hybrid multilevel inverters and electrical circuit analyses.



Bala gangi reddy received the B.Tech and M.Tech Degree from Jawaharlal Nehru Technological University, College of